

1. a)

$$x(t) = 30 \cos(\theta(t)), \quad \theta(t) = 10^6 t + 4 \sin(15t)$$

$$f_i(t) = \frac{1}{2\pi} \frac{d}{dt} \theta(t) = \boxed{\frac{1}{2\pi} (10^6 + 60 \cos(15t))}$$

using Carson's rule (other estimates are OK too):

$$BW \approx 2 \left(\frac{f_\Delta}{W} + 1 \right) W = 2(f_\Delta + W) \quad (W \text{ is the bandwidth of the message, } \frac{15}{2\pi} \text{ Hz})$$

$$= 2 \left(\frac{60}{2\pi} + \frac{15}{2\pi} \right) = \boxed{23.87 \text{ Hz}}$$

1. b)

$$y(t) = 30 \cos(5\theta(t))$$

$$f_i(t) = \frac{1}{2\pi} \frac{d}{dt} 5\theta(t) = \boxed{\frac{5}{2\pi} (10^6 + 60 \cos(15t))}$$

$$BW \approx 2(f_\Delta + W) = 2 \left(\frac{5 \cdot 60}{2\pi} + \frac{15}{2\pi} \right) = \boxed{100 \text{ Hz}}$$

1. c)

$$k_{fm}(t) = f_i - f_c = \boxed{\frac{60}{2\pi} \cos(15t)}$$

I realized later that this is ambiguous, so it's OK if you did it for $y(t)$ instead:

$$k_{fm}(t) = f_i - f_c = \boxed{\frac{300}{2\pi} \cos(15t)}$$

1. d)

$$k_{pm}(t) = \int (f_i - f_c) dt = \boxed{\frac{4}{2\pi} \sin(15t)}$$

or

$$k_{pm}(t) = \boxed{\frac{20}{2\pi} \sin(15t)}$$

1. e)

No units are given, so we just find the (rms)². This is because $f_c \gg f_\Delta$, so the signal(s) is essentially just sinusoidal.

$$p_x = p_y = \boxed{\frac{30^2}{2}}$$

2.a)

$$f_{c1} = 3 \text{ MHz}, f_{c2} = 7 \cdot f_{c1} = 21 \text{ MHz}, f_{c3} = 11 \cdot f_{c2} = 231 \text{ MHz}$$

2.b)

$$f_{d1} = 500 \text{ kHz}, f_{d2} = 7 \cdot f_{d1} = 3.5 \text{ MHz}, f_{d3} = 11 \cdot f_{d2} = 38.5 \text{ MHz}$$

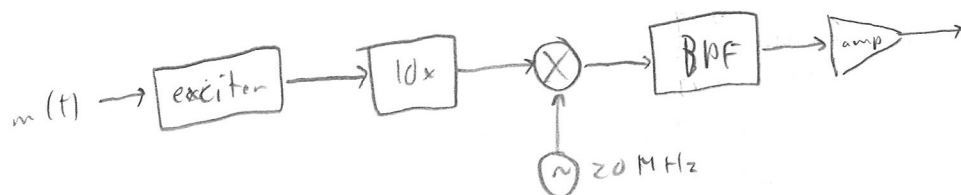
Note: I accidentally typed "kHz" instead of just "Hz". f_d is probably too big compared to f_c to be realistic. We can still go through the calculations though.

2.c)

$$\begin{aligned} BW_{x1} &= 2(f_{d1} + W) = 2(500 \text{ k} + 20 \text{ k}) = 1040 \text{ kHz} \\ BW_{x2} &= 2(f_{d2} + W) = 2(3.5 \text{ M} + 20 \text{ k}) = 7040 \text{ kHz} \\ BW_{x3} &= 2(f_{d3} + W) = 2(38.5 \text{ M} + 20 \text{ k}) = 77040 \text{ kHz} \end{aligned}$$

3.a)

$f_{d1} = 10 \text{ kHz}, f_{d0} = 100 \text{ kHz} \rightarrow$ we need a $10\times$ frequency multiplier
 $f_{c1} = 8 \text{ MHz}, f_{c0} = 60 \text{ MHz}$ But $8 \cdot 10$ is 80, so we need to shift the signal from the multiplier down by 20 MHz.



The filter would have a center frequency of 60 MHz and a BW between $2(100 \text{ k} + 20) = 200 \text{ kHz}$ and $80 + 20 = 100 \text{ MHz}$. It would probably be easiest to use a BW near the high end.

4.)

$$W = 15 \text{ kHz}$$

$$BW = 2(f_d + W)$$

$f_d (\text{kHz})$	BW (kHz)
0.1	30.2
0.5	31
1	32
5	40
10	50
50	130
100	230

5)

$$\frac{f_A}{W} = 5 \rightarrow f_{A \text{ music}} = 5 \cdot W_{\text{music}} = 5 \cdot 15 \text{ kHz} = 75 \text{ kHz}$$

$$BW_{\text{music}} = 2(75 + 15) = 180 \text{ kHz}$$

$$f_{A \text{ talk}} = 5 \cdot W_{\text{talk}} = 25 \text{ kHz}$$

$$BW_{\text{talk}} = 2(25 + 5) = 60 \text{ kHz}$$

$$\frac{BW_{\text{talk}}}{BW_{\text{music}}} = \frac{60}{180} = \boxed{33.3\%}$$

$$6.a) \quad \bar{X} = E\{X\} = \mu_X$$

$$\text{cov}(X, Y) = E\{(X - \bar{X})(Y - \bar{Y})\} = E\{XY - X\bar{Y} - Y\bar{X} + \bar{X}\bar{Y}\}$$

$$= \overline{XY} - \bar{X}\bar{Y} - \bar{Y}\bar{X} + \bar{X}\bar{Y} = \overline{XY} - \bar{X}\bar{Y}$$

If X and Y are independent, then $\overline{XY} = \bar{X}\bar{Y}$, so $\boxed{\text{cov}(X, Y) = 0}$

6.b)

$$Y = \alpha X + \beta \rightarrow \bar{Y} = E\{\alpha X + \beta\} = \alpha \bar{X} + \beta$$

$$\overline{XY} = E\{(\alpha X + \beta)X\} = E\{\alpha X^2 + \beta X\} = \alpha \overline{X^2} + \beta \bar{X}$$

$$\text{cov}(X, Y) = \alpha \overline{X^2} + \beta \bar{X} - \alpha \bar{X}^2 - \beta \bar{X} = \alpha \overline{X^2} - \alpha \bar{X}^2 = \boxed{\alpha (\overline{X^2} - \bar{X}^2) = \alpha \sigma_X^2}$$

7

$$\varepsilon^2 = E\{(Y - \tilde{Y})^2\}$$

$$(Y - \tilde{Y})^2 = (Y - \alpha X - \beta)^2 = Y^2 - \alpha XY - \beta Y - \alpha XY + \alpha^2 X^2 + \alpha \beta X - \beta Y + \alpha \beta X + \beta^2$$

$$= Y^2 - 2\alpha XY - 2\beta Y + \alpha^2 X^2 + 2\alpha \beta X + \beta^2$$

$$\varepsilon^2 = \overline{Y^2} - 2\alpha \overline{XY} - 2\beta \bar{Y} + \alpha^2 \overline{X^2} + 2\alpha \beta \bar{X} + \beta^2$$

Find the critical points by finding where the first partial derivatives are 0.

$$\frac{\partial}{\partial \alpha} \varepsilon^2 = -2 \overline{XY} + 2\alpha \overline{X^2} + 2\beta \bar{X} = 0$$

$$\rightarrow \begin{bmatrix} 2\overline{X^2} & 2\bar{X} \\ 2\bar{X} & 2 \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \end{bmatrix} = \begin{bmatrix} 2\overline{XY} \\ 2\bar{Y} \end{bmatrix}$$

$$\frac{\partial}{\partial \beta} \varepsilon^2 = -2\bar{Y} + 2\alpha \bar{X} + 2\beta = 0$$

7 (cont.)

HW5

We can use row reduction, (Gaussian elimination), matrix inverse, etc. to find α and β .

I used Mathematica to do it for me.

$$\alpha = \frac{\overline{XY} - \bar{X}\bar{Y}}{\overline{X^2} - \bar{X}^2} = \frac{\text{Cov}(X, Y)}{\sigma_x^2}$$

$$\beta = - \frac{\bar{X} \overline{XY} - \overline{X^2} \bar{Y}}{\overline{X^2} - \bar{X}^2}$$

with a bit of effort, β can be simplified

$$\bar{X} \overline{XY} - \overline{X^2} \bar{Y} = \bar{X} (\overline{XY} - \bar{X}\bar{Y} + \bar{X}\bar{Y}) - \overline{X^2} \bar{Y} = \bar{X} (\overline{XY} - \bar{X}\bar{Y}) + \bar{X}^2 \bar{Y} - \overline{X^2} \bar{Y}$$

$$= \bar{X} \text{Cov}(X, Y) - \bar{Y} (\overline{X^2} - \bar{X}^2) = \bar{X} \text{Cov}(X, Y) - \bar{Y} \sigma_x^2$$

$$\beta = - \frac{\bar{X} \text{Cov}(X, Y) - \bar{Y} \sigma_x^2}{\sigma_x^2} = -(\bar{X} \alpha - \bar{Y}) = \boxed{\bar{Y} - \alpha \bar{X}}$$

Finally, we need to check that the critical point is actually a minimum with the 2nd partial derivative test.

$$\begin{aligned} H &= \left(\frac{\partial^2}{\partial \alpha^2} \epsilon^2 \right) \left(\frac{\partial^2}{\partial \beta^2} \epsilon^2 \right) - \left(\frac{\partial^2}{\partial \alpha \partial \beta} \epsilon^2 \right)^2 \\ &= (2\overline{X^2}) (2) - (2\bar{X})^2 = 4(\overline{X^2} - \bar{X}^2) = 4\sigma_x^2 > 0 \end{aligned}$$

$$\frac{\partial^2}{\partial \alpha^2} \epsilon^2 = 2\overline{X^2} > 0 \quad \text{so the point we found must be a minimum.}$$